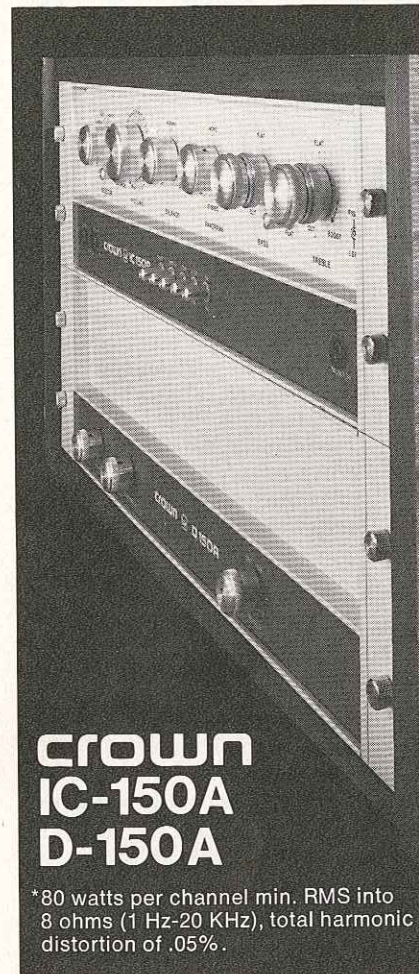


# Hi-fi Harmony



## Crown IC-150A D-150A

\*80 watts per channel min. RMS into 8 ohms (1 Hz-20 KHz), total harmonic distortion of .05%.

In any great vocal group, each singer could make it on his own. Together, they're beautiful.

That's how we designed the Crown IC-150A pre-amp and the Crown D-150A power amp.

The IC-150A provides precise control with minimum distortion. The D-150A provides reliable power\* to drive almost any stereo speakers. Each one could be a superb addition to your system.

But together we think they're incomparably beautiful, whether you think of beauty as powerful, clean sound; or honest, uncluttered appearance; or simple, accurate control.

If the receiver or integrated amp in your current system seems somehow limited, please consider seriously the Crown IC-150A/D-150A combination. You'll be pleased at a new quality of sound, and surprised at a real bargain.

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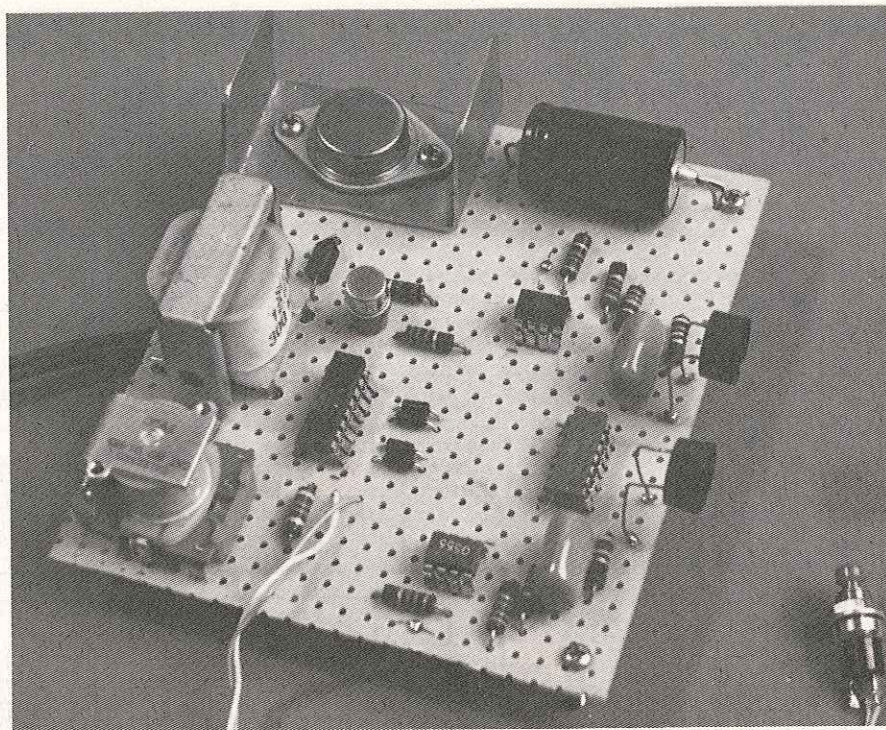


Fig. 4. The prototype of the photoelectric sensor was assembled entirely on perforated board. LDR's are on right side of board. CLEAR switch is connected by twisted leads.

advisable to use sockets for the IC's and transistor. Note also the need for a heat sink with IC5.

Light-dependent resistors *LDR1* and *LDR2* should be mounted about 1" (2.54 cm) apart so that a single light beam will suffice for both. If you are using extra UP/DOWN counting inputs, mount their LDR's on a small piece of perforated board. Cut holes in a small box to allow light-beam access to the LDR pair, mount the LDR board inside the box, and interconnect this assembly with the main board via twisted cable.

The box that houses the main circuit board should be large enough to accommodate the main circuit board, a chassis-mounting ac receptacle, and CLEAR pushbutton switch *S1*. Drill holes in the box as required to mount all components in place and to provide light-beam access to *LDR1* and *LDR2*. Mount *S1* and the receptacle in their respective holes. You can use ordinary hookup wire to interconnect *S1* with the circuit board, but it is advisable that you use a length of regular line cord to interconnect the relay contacts and the receptacle. Slip the free end of the line cord through a rubber-grommet-lined hole in the case and solder it to the appropriate points in the circuit. A sheet of insulating plastic should be placed between the box and the bottom of the board before the latter is finally bolted down. This will obviate the possibility of the live ac on

the primary side of *T1* from shorting out against the metal box.

**Testing the Circuit.** Plug the system's line cord into a convenient ac receptacle and direct a beam of light onto *LDR1* and *LDR2*. The relay should immediately energize. Depressing the CLEAR button (*S1*) should cause the relay to immediately deenergize. Now, block the light beam by passing your hand in front of first *LDR1* and then *LDR2*. The relay should again energize. With the relay still energized, passing your hand in front of first *LDR2* and then *LDR1* should cause *K1* to drop out.

Pass your hand several times from *LDR1* to *LDR2*. The relay should immediately energize on the first pass and remain energized with each successive pass. Now pass your hand an equal number of times from *LDR2* to *LDR1*. The relay should remain energized for all but the last pass. On the last pass, the relay should deenergize. This procedure checks the up and down counting operation of the circuit.

The relay specified for *K1* in the Parts List can safely handle up to 3 amperes of current. If you wish to operate a device that requires a greater amount of current, you will have to substitute a low-voltage, low-current relay whose contacts can handle the current drain. Alternatively, you can use the specified relay to drive a higher-current relay. ◇

POPULAR ELECTRONICS

# HOW TO FULLY DEBOUNCE LOW-COST KEYBOARDS



## Inexpensive approach to properly interfacing calculator keyboards to microcomputers

BY RALPH TENNY

**KEYBOARDS** for home-brew calculators or even minicomputers are available from many sources (such as those advertising in this magazine). Unfortunately, most of these keyboards lack two important features—contact debouncing and key encoding. Without debouncing, each key closure can produce multiple signals, while encoding makes it possible to determine just which key has been operated.

There are many different types of debouncing circuits, but most of them are applicable to just one contact. Debouncing a full keyboard can be very expensive. Also, when debouncing a keyboard, encoding is very complex since each key closure must result in a unique code output.

A simple, low-cost way to overcome these problems is shown in the circuit in

Fig. 1. It produces a binary-coded output, fully debounced, from any low-cost 16-key board. The keys are labelled in hexadecimal so that the board can communicate with a microprocessor.

Four 8-input NAND gates (*IC1* through *IC4*) encode all the keys except 0. All of these gates are held high by resistors *R4* through *R19*. When a key is depressed (except 0), that input is brought to ground and its associated NAND gate output goes high. For example, if key D is depressed, the outputs of *IC1*, *IC3*, and *IC4* go high to produce 1101, the hex code for D.

Because it is necessary to know when any key, including 0 is depressed, OR gates *IC6A* and *IC6B* detect the presence of any key closure. When a 0 comes on, *IC5A* passes the signal to gate *IC6B*.

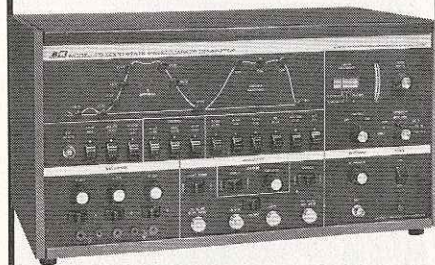
Although we can now detect all key closures and encode them on the 1-2-4-8 lines, contact bounce remains a problem. The waveforms in Fig. 2 show what happens when a key is depressed and bounces one time. The output of *IC6B* is waveform M, which drives the *R1-C1* combination to produce waveform N. When the key stops bouncing, *C1* can be charged up enough to cause *Q1* to fire. Resistor *R1* is low enough so that *Q1* latches in and stays on. As a result, point P is held low, and the two sections of *IC5* produce high or low Data Ready signals. The latter is indicated by a rule over the words in Fig. 1.

When *Q1* latches in, it provides some protection against multiple-key closures. If a second key is depressed after *Q1* fires, the output code will be correct but no Data Ready strobe will be produced.

JANUARY 1977

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# Fast, Fool-proof TV Alignment



## B&K-PRECISION SWEEP/MARKER GENERATOR

Model 415, \$485

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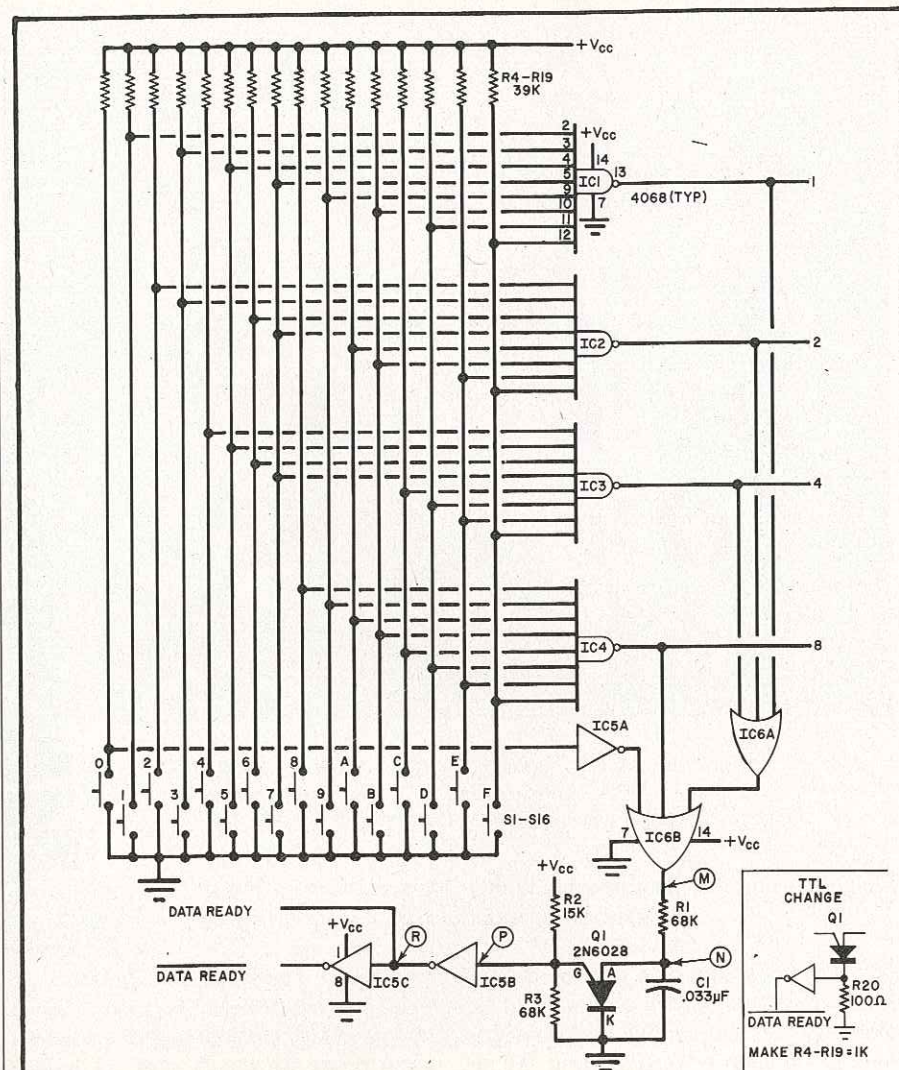


Fig. 1. The circuit produces a 1-2-4-8 binary code and provides a debounced data-ready strobe.

### PARTS LIST

C1—0.033- $\mu$ F, 15-V capacitor	Q1—2N6028 programmable unijunction transistor
IC1-IC4—CD4068 8-input NAND gate (for TTL, use 7430)	R1, R3—68,000-ohm, 1/4-watt resistor
IC5—CD4049 inverting hex buffer (for TTL, use 7404)	R2—15,000-ohm, 1/4-watt resistor
IC6—CD4075 triple 3-input OR gate (for TTL, use 7432)	R4-R19—39,000-ohm, 1/4-watt resistor
	R20—100-ohm, 1/4-watt resistor (TTL only)
	S1-S16—Normally open switch/keyboard

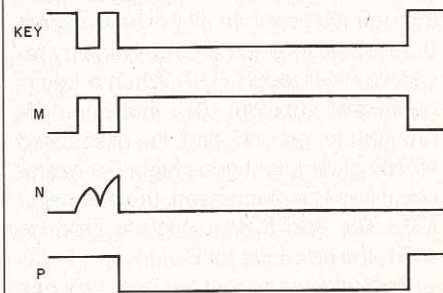


Fig. 2. Waveforms show how bouncing key can produce a clean single output.

The UJT will reset only after all the keys are released, and the output of IC6B returns to zero. If a second key is closed within a few milliseconds (while the first key is still bouncing), an erroneous output can be produced.

The keyboard can be battery powered if the CMOS devices are used. If you are going to drive TTL logic with this adaptor, change the IC's to their TTL counterparts (see Parts List), change the values of R4 through R19 to 1000 ohms, and add a 100-ohm resistor in the cathode of Q1. The Data Ready signal is then available as shown in Fig. 1. ◇

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